

ECSE 4962 Control Systems Design

A Brief Tutorial on Control Design

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<http://www.cat.rpi.edu/~wen/ECSE4962S04/>

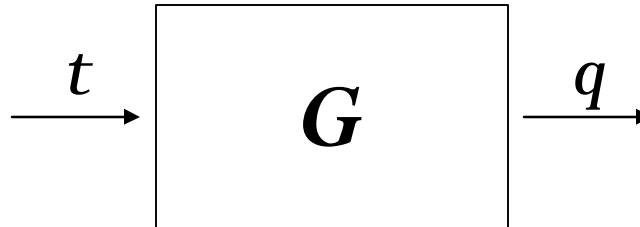
Don't Wait Until The Last Minute!

- **You got to have a model to work with by now:**
 - **At least a model based on estimated mass/inertia and your motors and gears (very few groups have this in the proposal)**
 - **You should very soon have an identified model.**
 - **If you don't, you must seek help!**
- **Once you have a model, start control design – in simulation. Don't just tweak the controller with the experiment without some simulation based analysis first!**

Today

- Review basic control design
- PID tuning

LTI System Characterization

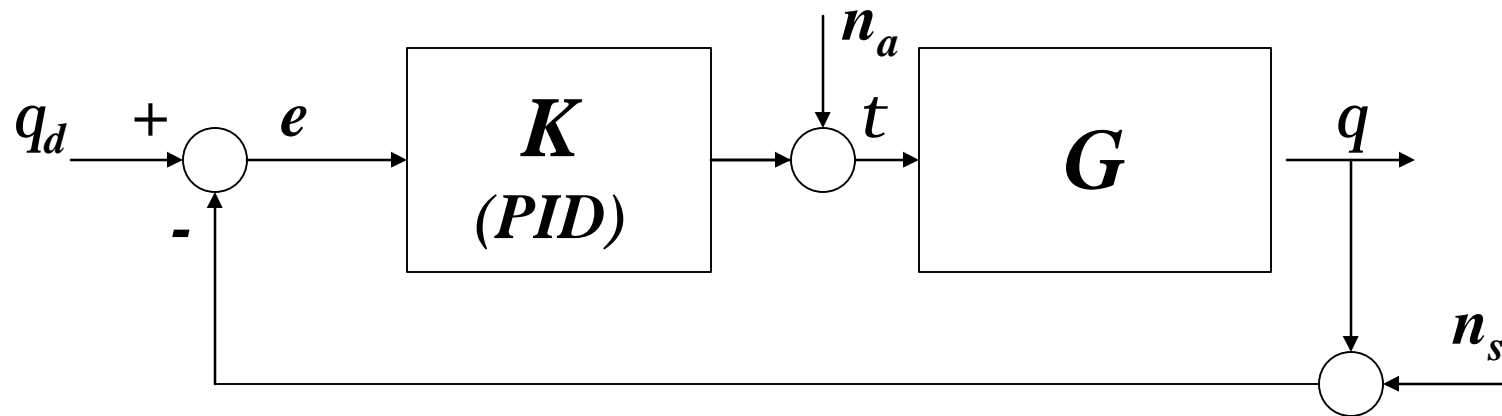


We'll approximate each axis as an independent single-input/single-output (SISO) system: $q = G t$

Characterization:

- poles and zeros `zpk(G)` ;
- frequency response `bode(G)`
- step response `step(G)`
- steady state value `evalfr(G, 0)`

Closed Loop System



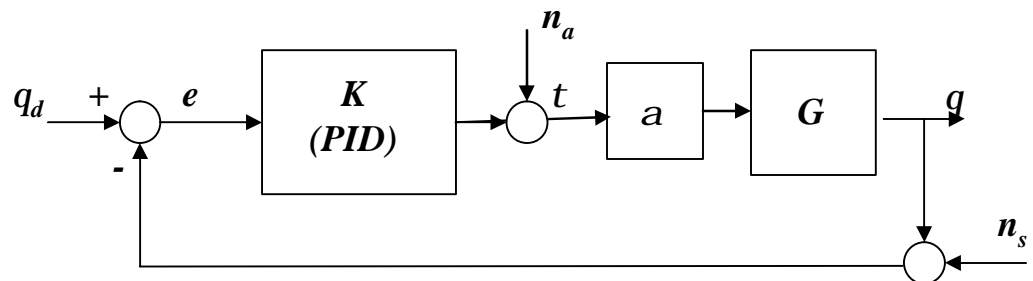
Relevant transfer functions:

$$q = \frac{KG}{1+KG} (q_d - n_s) + \frac{G}{1+KG} n_a$$

$$e = \frac{1}{1+KG} q_d + \frac{KG}{1+KG} n_s - \frac{G}{1+KG} n_a$$

Design Objective

- **Stability:** closed loop poles must all be in left half plane.
- **Performance:**
 - Step response has small overshoot and small settling time.
 - Small steady state error
- **Disturbance Rejection:** Effect of sensor and actuator noises small on q
- **Robustness:** How large an uncertainty a can be tolerated (in terms of stability)?



Design Objective

- **Stability: closed loop poles must all be in left half plane.**

closed loop poles = roots of $(1 + KG)$

Design Objective

- **Performance:**
 - **Step response has small overshoot and small settling time.**

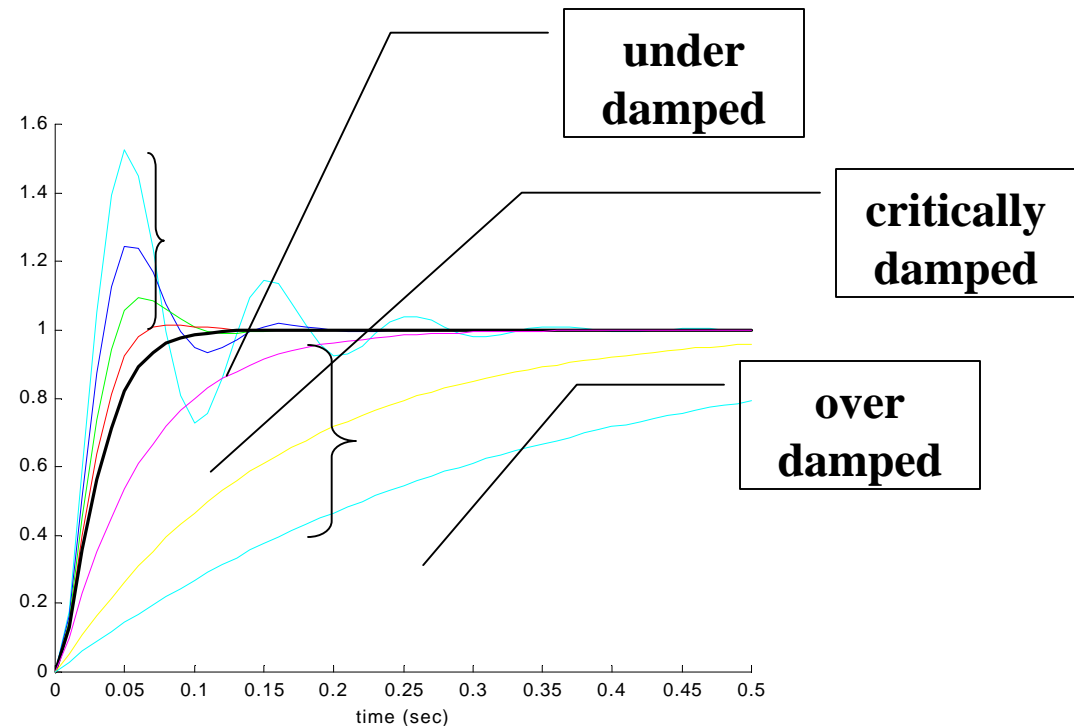
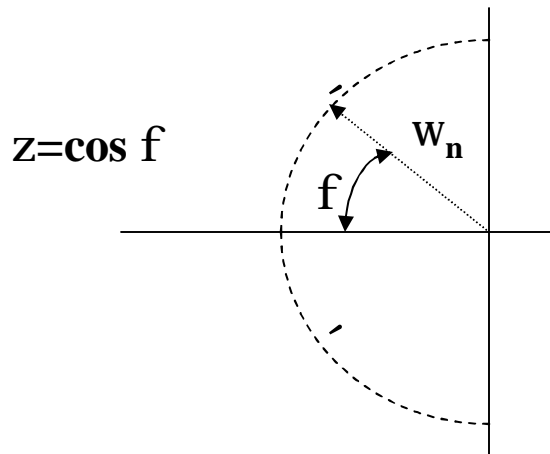
closed loop transfer function: $G_{CL} = \frac{GK}{1 + GK}$

w_n sufficiently large, z sufficiently close to 1

Step Response

- Intuition based on second order system with no zero

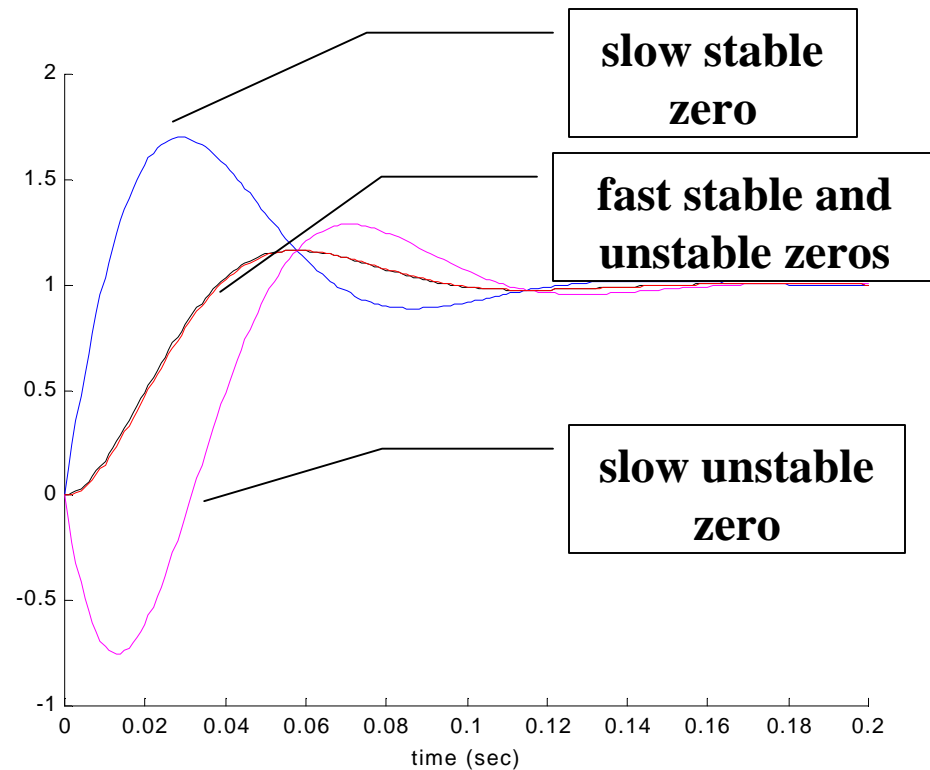
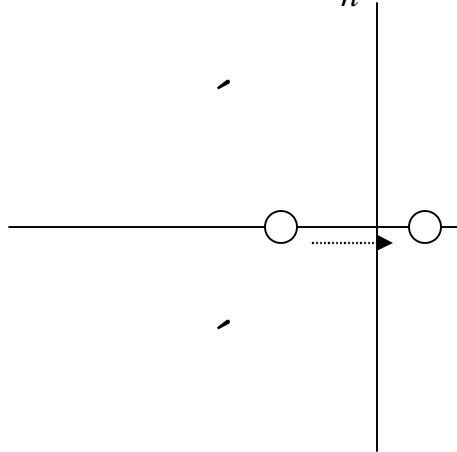
$$H(s) = \frac{w_n^2}{s^2 + 2zw_n s + w_n^2}$$



Effect of Zero

- Zero within system bandwidth strongly affects response
- Stable zero increases overshoot, unstable zero gives rise to undershoot

$$H(s) = \frac{w_n^2 (s/a + 1)}{s^2 + 2\zeta w_n s + w_n^2}$$



Design Objective

- **Performance:**
 - **Small steady state error**

$$\mathbf{q}_{ss} = G_{CL}(0)(\mathbf{q}_d - n_{s,DC}) + \frac{G(0)}{1 + G(0)K(0)} n_{a,DC}$$
$$G_{CL} = \frac{GK}{1 + GK}$$

You want this
close to 1

You want this
close to 0

To reduce steady state error:

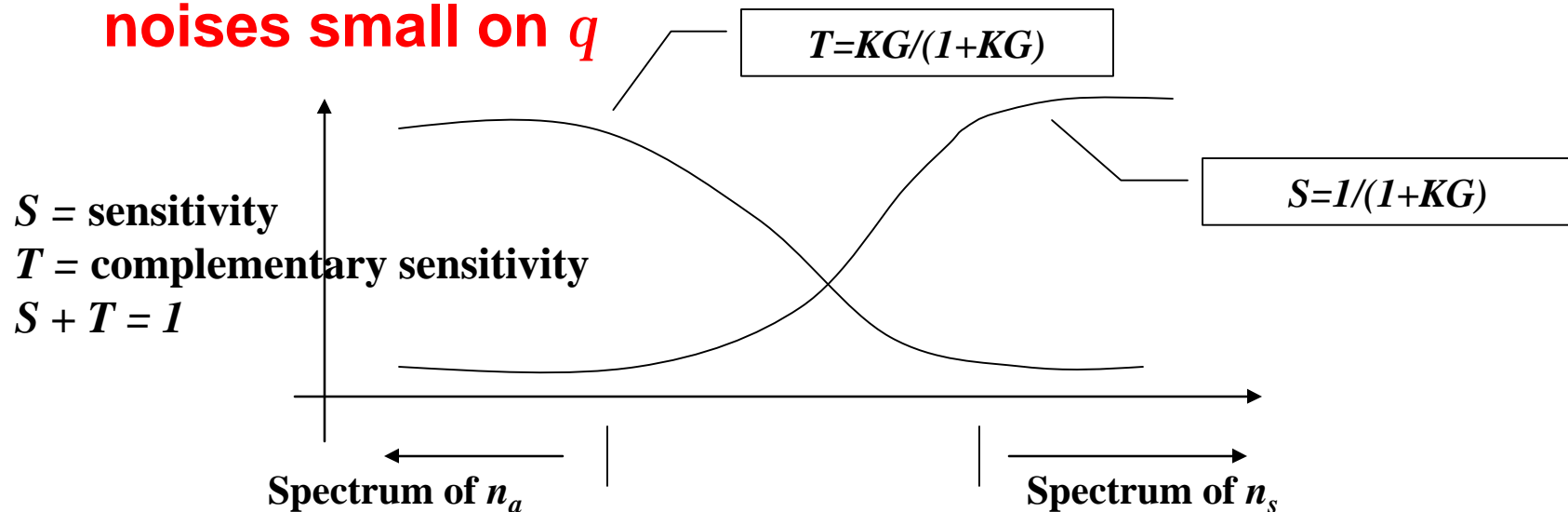
- Cancel n_a if possible
- High DC loop gain $G(0)K(0) \rightarrow$ Integral control

Design Objective

$$e = \frac{1}{1+KG} \mathbf{q}_d + \frac{KG}{1+KG} n_s - \frac{G}{1+KG} n_a$$

Design $KG/(1+KG)$ small over the spectrum of n_s and $G/(1+KG)$ small over the spectrum of n_a

- Disturbance Rejection: Effect of sensor and actuator noises small on q**



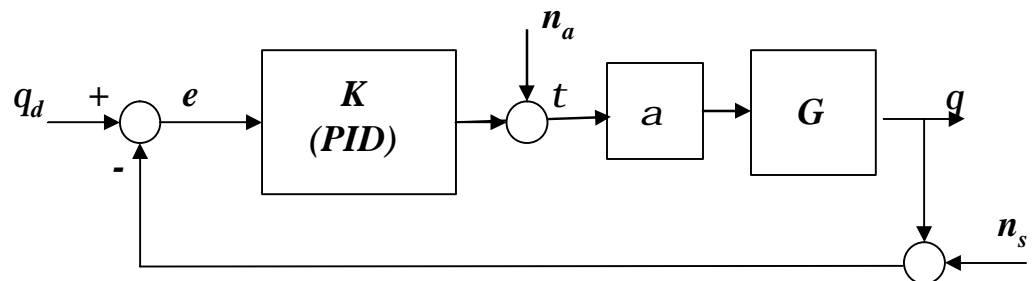
Design Objective

Gain margin: if a is a real number, how far can a be different from 1 before the closed loop system becomes unstable?

Phase margin: if $a = \exp(jq)$ (a pure phase shift), how large can q be before the closed loop system becomes unstable?

Usually expressed in terms of dB:
 $a \in [.5, 2]$ means GM=6dB

- Robustness:** How large an uncertainty a can be tolerated (in terms of stability)?

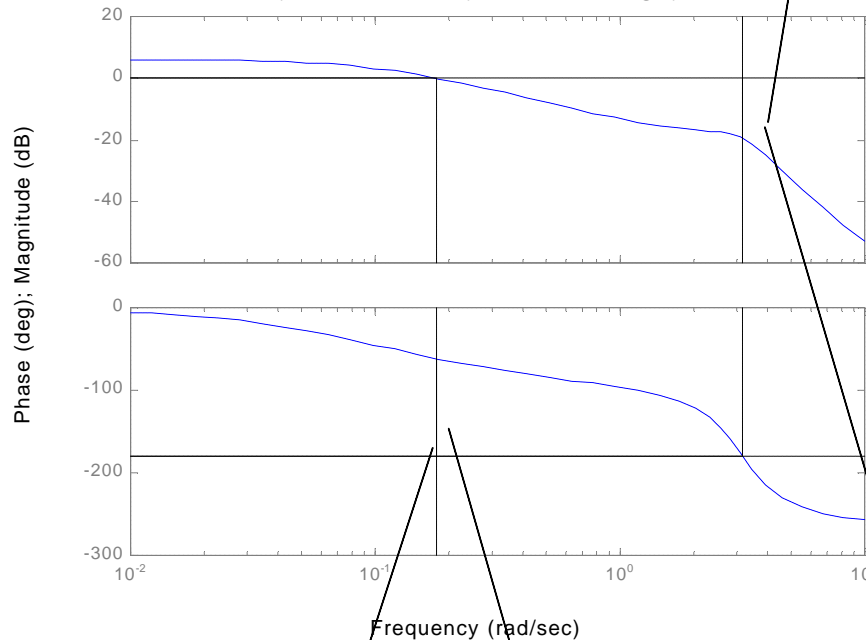


Robustness (Gain/Phase Margins)

a is nominally 1

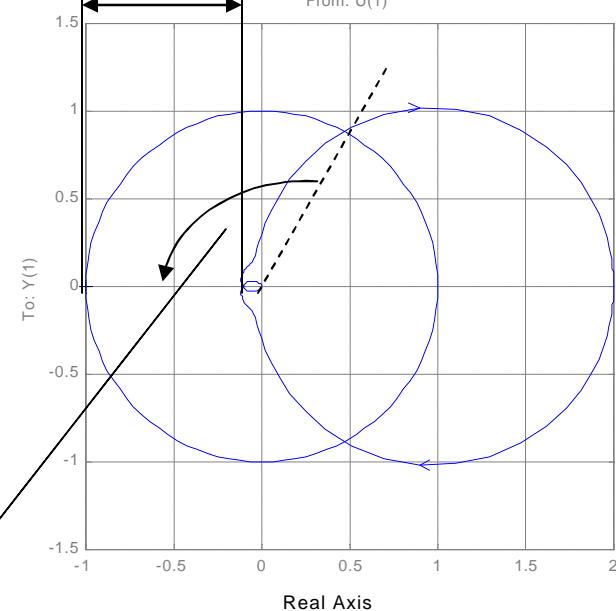
Bode Diagrams

Gm=19.554 dB (at 3.1623 rad/sec), Pm=117.94 deg. (at 0.17724 rad/sec)



gain
margin

Nyquist Diagrams
From: U(1)



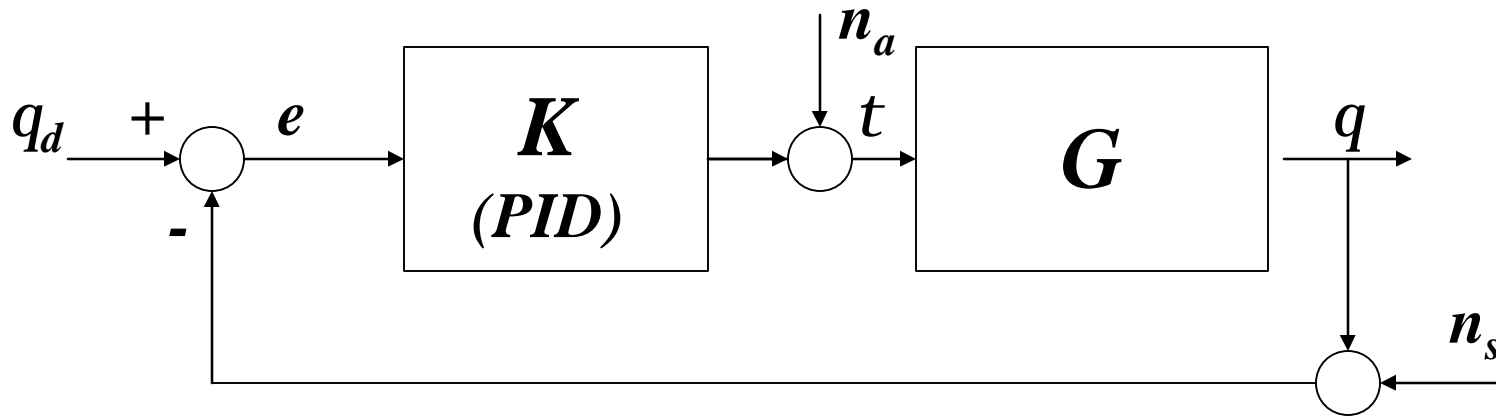
Add phase
lead for phase
stabilization

phase
margin

reduce gain for
gain stabilization

use `margin` command in MATLAB

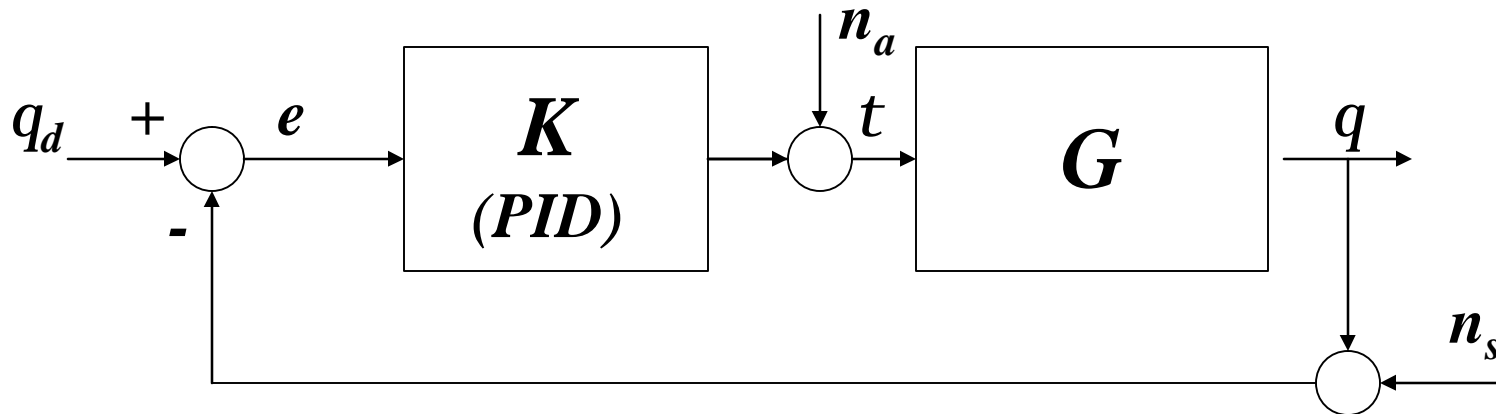
PID Control



$$K(s) = k_P + k_I / s + k_D s$$

When does it work?

PID Control



$$K(s) = k_P + k_I / s + k_D s$$

Works well when G is a 2nd order system.

PID Control

Consider $G(s)=1/s^2$: closed loop characteristic polynomial is $s(s^2 + K_D s + K_P) + K_I$

For small K_I , K_D governs the damping and K_P governs the undamped natural frequency w_o

For $K_I > 0$, DC loop gain is infinite, therefore, zero steady state error.

Gain Tuning

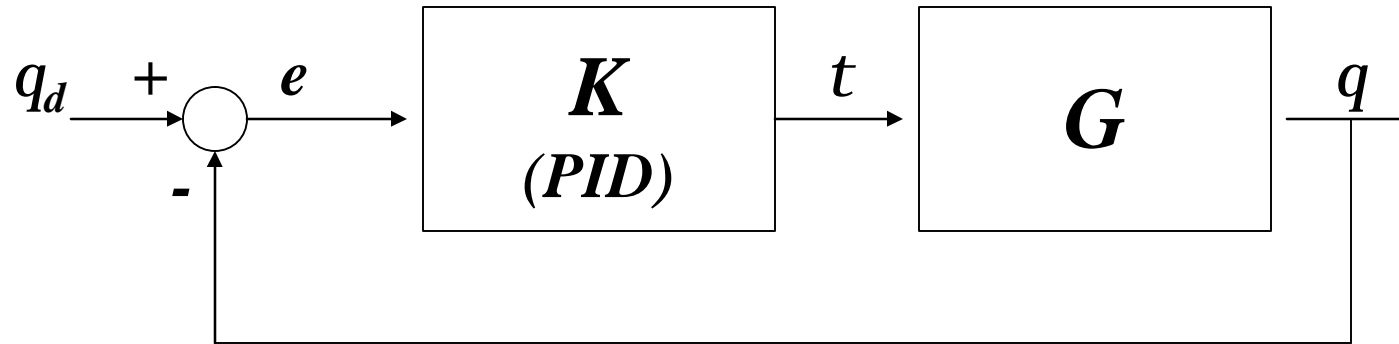
- **Intuition:**

- P gain increases speed of response but also increases overshoot
- D gain reduces overshoot but decreases speed of response
- I gain reduces steady state error but can reduce speed of response and lead to instability

- **Strategy:**

- Tune PD gain until desired transient response is obtained
- Increase I gain until convergence to steady state is satisfactory.
- Retune PD gains (increase) if necessary.

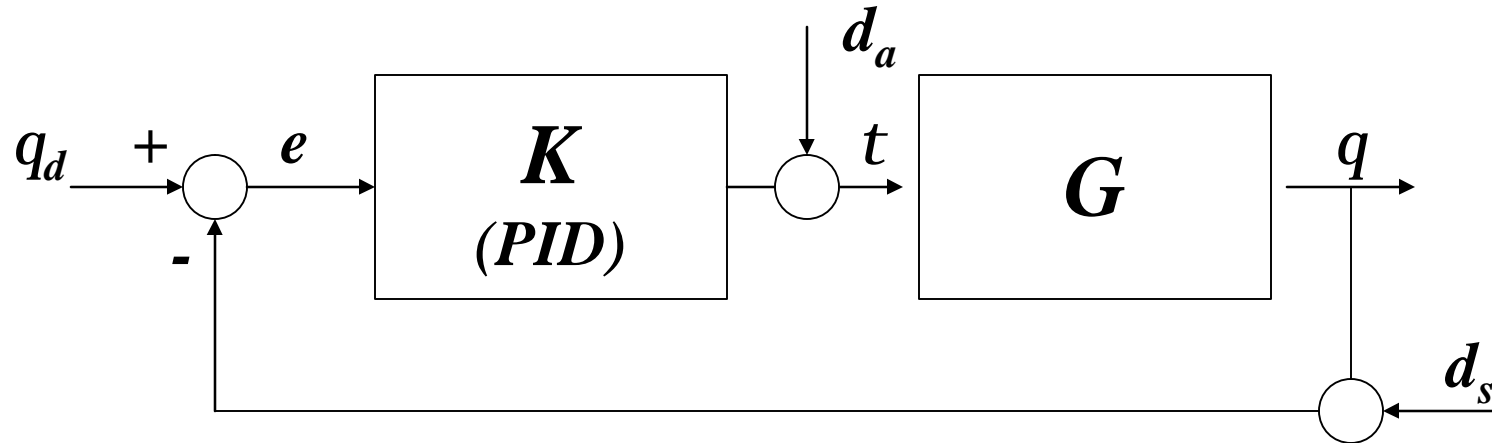
Frequency Domain Considerations



Adjust PID gains to achieve

- **good tracking over the desired bandwidth (transfer function from q_d to q is close to 0dB)**

Disturbance Rejection



Adjust controller (and possibly add more filtering in K , but must be careful to preserve stability and dynamical response) so the frequency gain is small over the disturbance frequency.

Next Week

- **More on control design.**

Tomorrow at 6pm in CII 2037

- **Group 1: 6pm, Group 2: 6:15pm, Group 3: 6:30pm, Group 4: 6:45pm, Group 5: 7:00pm, Group 6: 7:15pm, Group 7: 7:30pm.**
- **Prepare to discuss the progress of your project.**