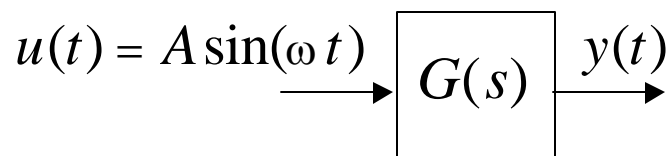


Today (10/16/01)

- **Today**
 - **Collect Project 2**
 - **Hand out Project 3**
 - **Frequency response**
 - **Ref. 6.1**
- **Reading Assignment: 6.9**

Frequency Response

- Consider a stable system driven by a sinusoidal input (like in Project 1).



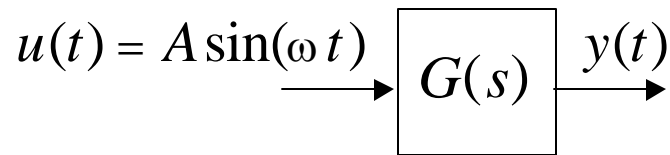
After the initial transient dies out,

$$y(t) = |G(j\omega_o)| U_o \sin(\omega_o t + \angle G(j\omega_o))$$

$$\left. \begin{array}{l} \text{magnitude of } G: |G(j\omega)| \\ \text{phase of } G: \angle G(j\omega) \end{array} \right\} \text{Frequency response of } G$$

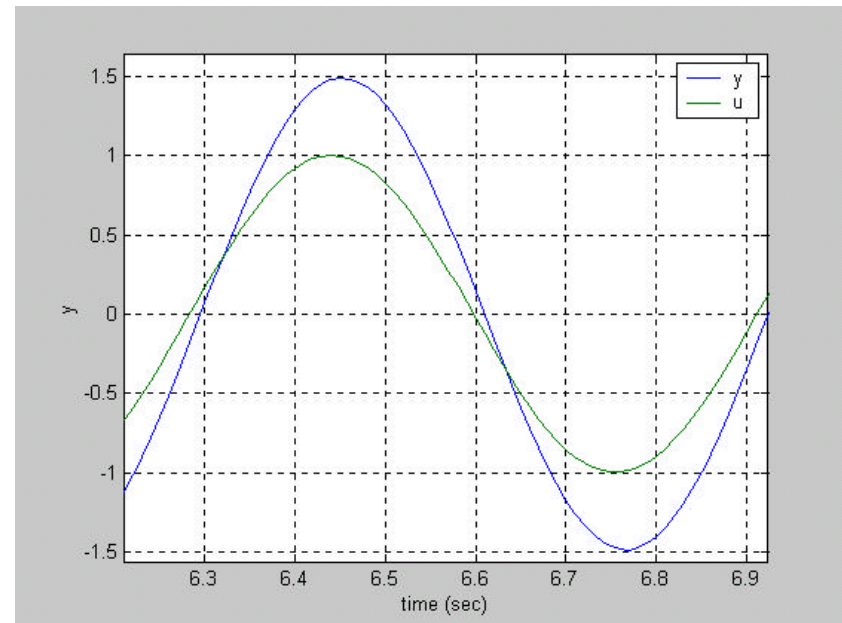
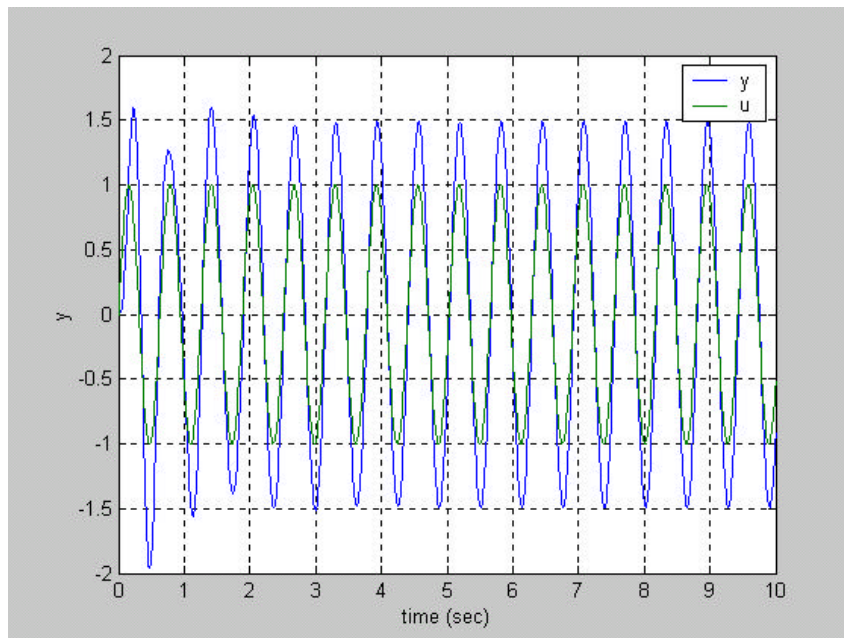
- Let $u(j\omega)$ be the Fourier transform of $u(t)$ (spectrum of u). $G(j\omega)$ can be considered as the gain scaling and phase shift of each sinusoidal component of u .

Example



$$\text{Let } G(s) = \frac{300}{s^2 + 2.427s + 300}$$

$$G(10j) = 1.49 \exp(-j0.12)$$



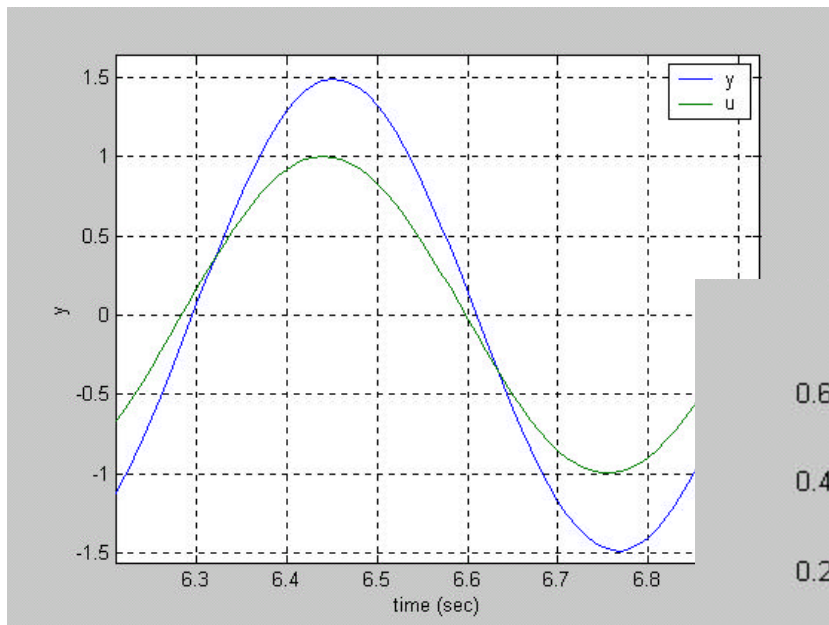
Example

$$G(10j) = 1.49\exp(-j0.12)$$

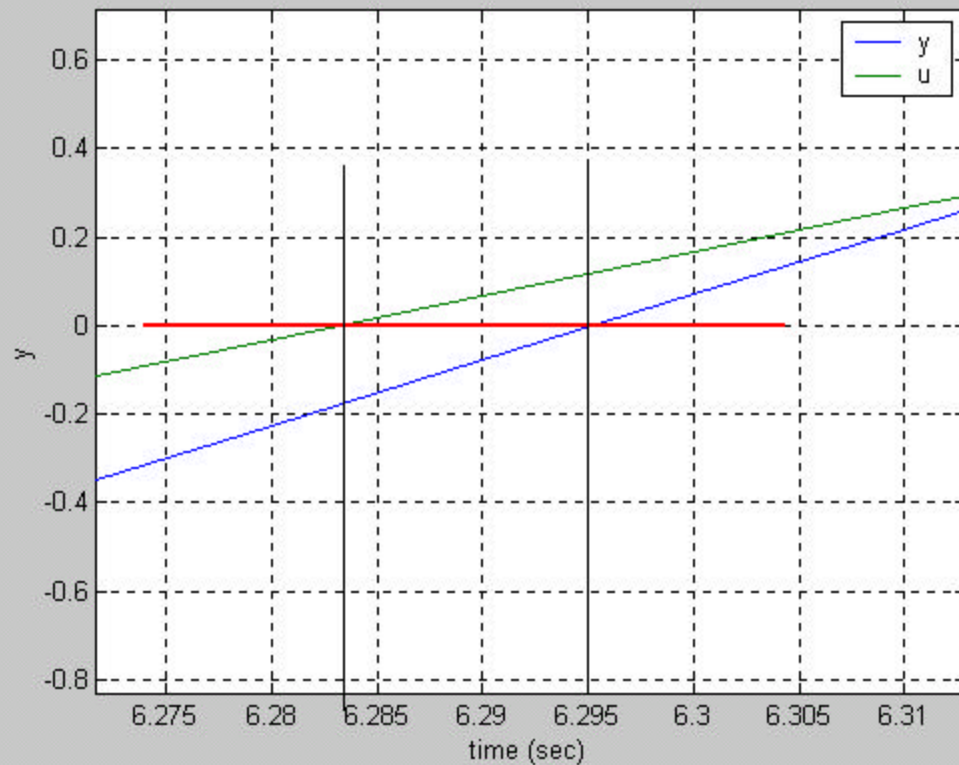
$$\sin(10(t - \phi / 10))$$

$\phi / 10 =$ separation of zero crossings

$$\phi / 10 = 0.012 \text{ rad or } \phi = 0.12 \text{ rad}$$



peak of y / peak of $u \approx 1.5$



Example

```
a=2.4274;  
G=tf(1,[1 a 0]);  
G1=feedback(G*300,1);  
G1_10=evalfr(G1,10*j);  
disp(sprintf('gain of G(10j) = %g',abs(G1_10)));  
disp(sprintf('phase of G(10j) = %g',angle(G1_10)));  
t=(0:.01:10);  
u=sin(10*t);  
y=lsim(G1,u,t);  
figure(1);plot(t,y,t,u);grid;  
legend('y','u');xlabel('time (sec)');ylabel('y');
```

Bode Plot

- **Gain plot is represented in decibels (dB) of $|G(j\omega)|$ vs. $\log_{10} \omega$: $|G|_{dB} = 20\log|G|$**
- **Phase plot is $\angle(G(j\omega))$ vs. $\log_{10} \omega$ (in degree or radian)**

$$G_1 G_2 = |G_1| |G_2| \exp(j\angle(G_1)) \exp(j\angle(G_2))$$

$$|G_1 G_2| = |G_1| |G_2| \quad \exp(j\angle(G_1 G_2)) = \exp(j(\angle(G_1) + \angle(G_2)))$$

- **Both gain/phase plots are additive**

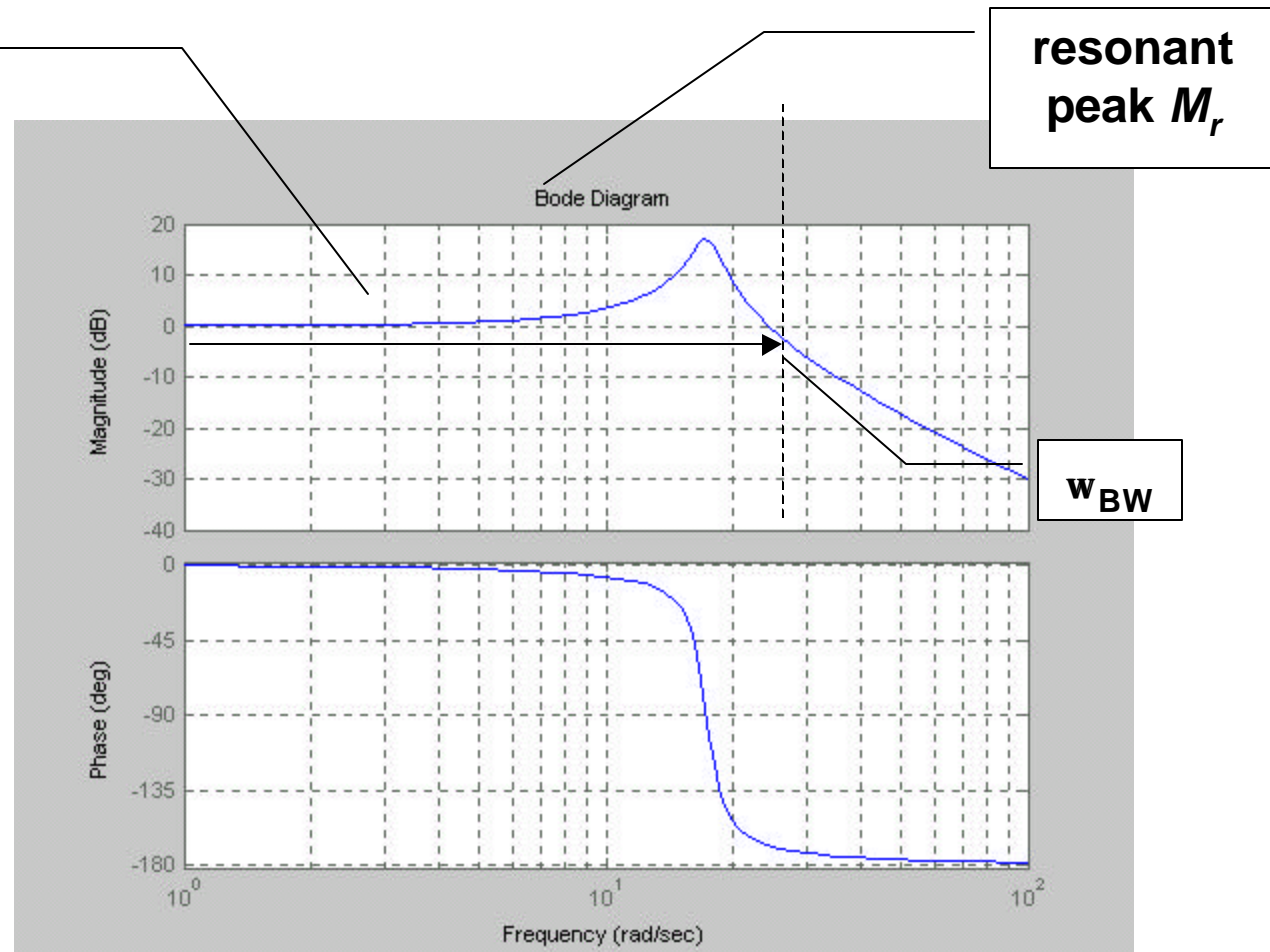
$$|G_1 G_2|_{dB} = 20\log|G_1 G_2| = 20\log|G_1| + 20\log|G_2| = |G_1|_{dB} + |G_2|_{dB}$$

$$\angle(G_1 G_2) = \angle(G_1) + \angle(G_2)$$

Bode Plot

- To visualize complete frequency spectrum of $G(s)$, use MATLAB command `bode(G)`

bandwidth:
frequency range over which input sinusoids can be satisfactorily tracked. Typically chosen as -3dB drop from DC value.



Bode Plot by hand

Consider Bode plot of loop gain $KG(s)$: 

$$KG(s) = K \frac{(s - z_1)(s - z_2) \dots}{(s - p_1)(s - p_2) \dots} = K_o \frac{\left(\frac{s}{z_1} - 1 \right) \left(\frac{s}{z_2} - 1 \right) \dots}{\left(\frac{s}{p_1} - 1 \right) \left(\frac{s}{p_2} - 1 \right) \dots}$$

$$\begin{aligned} |KG(j\omega)|_{dB} &= |K_o|_{dB} + |j\omega / z_1 - 1|_{dB} + |j\omega / z_2 - 1|_{dB} + \dots \\ &\quad - |j\omega / p_1 - 1|_{dB} - |j\omega / p_2 - 1|_{dB} - \dots \end{aligned}$$

$$\begin{aligned} \angle (KG(j\omega)) &= \angle K_o + \angle (j\omega / z_1 - 1) + \angle (j\omega / z_2 - 1) + \dots \\ &\quad - \angle (j\omega / p_1 - 1) - \angle (j\omega / p_2 - 1) - \dots \end{aligned}$$

Bode Plot by hand

Different types of terms :

0. constants: c

1. zeros/poles at origin: s^n

2. real zeros/poles: $(s\tau + 1)^{\pm 1}$

3. complex zeros/poles: $\left[\left(\frac{s}{\omega_n} \right)^2 + 2\zeta \frac{s}{\omega_n} + 1 \right]^{\pm 1}$

We will develop rules for each term. For the final Bode plot, just add the individual gain and phase plots together.

Bode Plot by Hand

$$|c|_{dB} = 20\log_{10}|c| = \text{constant}$$

$$\angle(c) = \text{constant phase} = \begin{cases} 0^\circ & \text{if } c > 0 \\ 180^\circ & \text{if } c < 0 \end{cases}$$

$$|(j\omega)^n|_{dB} = n \cdot 20\log_{10}|\omega| \quad (20 \cdot n \text{ dB/decade})$$

$$\angle(j\omega)^n = n \cdot 90^\circ \quad (\text{constant phase})$$

Bode Plot by Hand

$j\omega\tau + 1$:

For $\omega\tau \ll 1$: $j\omega\tau + 1 \approx 1$ (0dB, 0 phase)

For $\omega\tau \gg 1$: $j\omega\tau + 1 \approx j\omega\tau$ (20dB/decade, 90°)

$\omega\tau = 1$: 3dB, 45°

Gain asymptotes: 0dB up to $\omega = 1/\tau$, 20dB/decade increase beyond

Phase asymptotes: Straight line approximation from 0° to 90° from 1 decade below $\omega = 1/\tau$ to 1 decade above $\omega = 1/\tau$, passing through 45° at $\omega = 1/\tau$.

Bode Plot by Hand

$$\beta(j\omega) = \left[\left(\frac{j\omega}{\omega_n} \right)^2 + 2\zeta \frac{j\omega}{\omega_n} + 1 \right]^{\pm 1} :$$

For $\omega \ll \omega_n$: $\beta(j\omega) \approx 1$ (0dB, 0 phase)

$$\text{For } \omega \gg \omega_n : \beta(j\omega) \approx \left[-\left(\frac{\omega}{\omega_n} \right)^2 \right]^{\pm 1} \quad (40\text{dB/decade}, \pm 180^\circ)$$

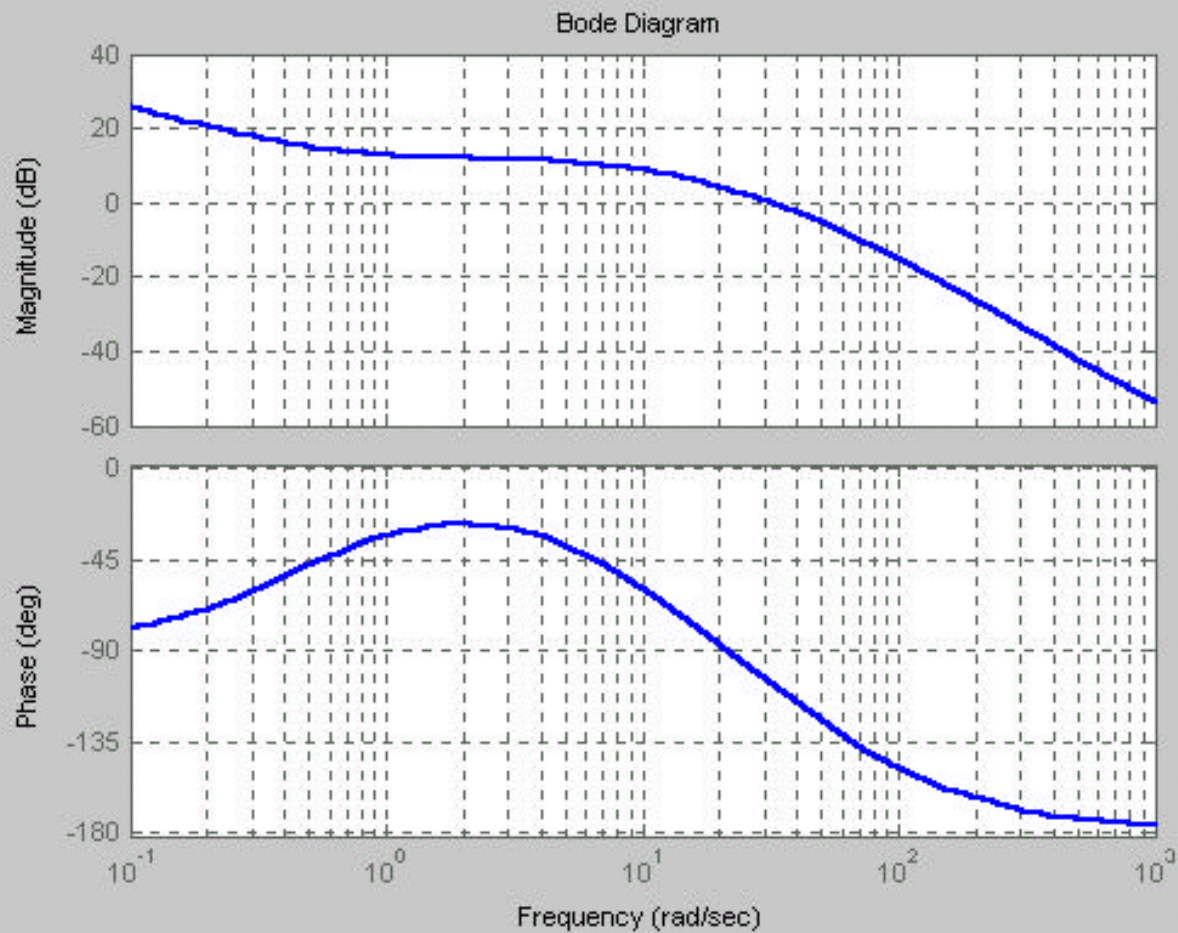
$\omega = \omega_n$: $\pm |2\zeta|$ (small $\zeta \Rightarrow$ large valley/peak), 90°

Gain asymptotes: 0dB up to $\omega = \omega_n$, 40dB/decade inc/dec beyond

Phase asymptotes: Straight line approximation from 0° to $\pm 180^\circ$ from 1 decade below $\omega = \omega_n$ to 1 decade above $\omega = \omega_n$, passing through 90° at $\omega = \omega_n$.

Example

$$KG(s) = \frac{2000(s + 0.5)}{s(s + 10)(s + 50)}$$



Exercise 8

Sketch by hand the Bode plot of

$$G(s)=1/(s^2+as)$$

Sketch by hand the Bode plot of lead

filter: $G(s)=K(Ts+1)/(\alpha Ts+1) \quad \alpha < 1$

Sketch by hand the Bode plot of lag filter:

$$G(s)=\alpha(Ts+1)/(\alpha Ts+1) \quad \alpha > 1$$